- Response to a step input is a primary consideration in discussions on control systems.
- What about responses to other types of input? Consider a sinusoidal input.
- For LTI systems, the response to a sinusoidal input is a sinusoid with the same frequency as the input sinusoid.

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## Frequency Response

- Develop techniques and tools for system design in the frequency domain.
  - Polar plots.
  - -Bode plots.
  - -Nyquist criterion and Nyquist plot.
- For today, we will see what frequency response is all about. Why do we need to use frequency response techniques?

We will look at a simple example of a polar plot.

- However, the sinusodial output has a different magnitude and phase in comparison to the sinusoidal input.
- Determine the magnitude and phase variations of the sinusoidal response when the frequency is varied.
- Define frequency domain specifications.

Relate time domain performance specifications to the frequency domain specifications.

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## Frequency Response

• What is the frequency response?

Frequency response - steady-state response of a system to a sinusoidal input signal.



the output has different a magnitude and phase compared to the input.

• Consider a system G(s) with unique poles  $p_1, \ldots, p_n$ .

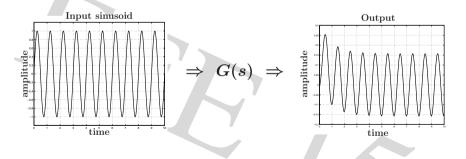
$$G(s) = \frac{(s - z_1) \dots (s - z_m)}{(s - p_1) \dots (s - p_n)}$$

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• Let the input 
$$r(t) = A \sin \omega t$$
.  
 $R(s) = \frac{A\omega}{s^2 + \omega^2}$   
• The output is  $Y(s) = G(s)R(s)$ .  
 $Y(s) = \frac{k_1}{s - p_1} + \dots + \frac{k_n}{s - p_n} + \frac{\alpha s + \beta}{s^2 + \omega^2}$   
 $y(t) = k_1 e^{p_1 t} + \dots + k_n e^{p_n t} + \mathscr{L}^{-1} \left\{ \frac{\alpha s + \beta}{s^2 + \omega^2} \right\}$ 

Detailed Look at the Frequency Response

• What does the actual response look like.



• Response to the sinusoid is composed of a transient part and steady-state part.

$$y(t) = y_{transient}(t) + y_{steady-state}(t)$$

©2003 M.C. Ramos UP EEE Institute • Using the final value theorem. (assuming poles  $p_1, \ldots, p_n$  are in the LHP),

$$\lim_{t o\infty}y(t)\ =\ \lim_{t o\infty}\mathscr{L}^{-1}\left\{rac{lpha s\ +\ eta}{s^2\ +\ \omega^2}
ight\}$$

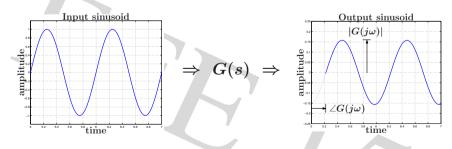
The steady-state output y(t) as  $t \to \infty$ ,  $y(t) = \frac{1}{\omega} |A\omega G(j\omega)| \sin(\omega t + \phi)$  $= A|G(j\omega)| \sin(\omega t + \phi)$  where  $\phi = \angle G(j\omega)$ 

• Output depends on the magnitude and phase of  $G(j\omega)$ .

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## Detailed Look at the Frequency Response

• Steady-state response.



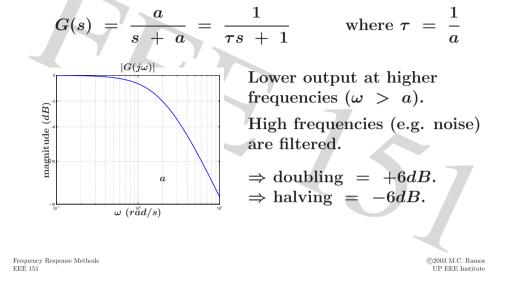
• For the input signal  $\sin(\omega t)$ , the output has a frequency  $\omega$ , a peak value of  $|G(j\omega)|$ and is shifted by  $\angle G(j\omega)$  with respect to the input. • Magnitude  $|G(j\omega)|$  expressed in dB (decibels). Logarithmic gain  $\stackrel{\triangle}{=} 20 \log |G(j\omega)|$ .

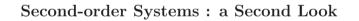
• Bandwidth  $\stackrel{\triangle}{=}$  frequency where magnitude drops by  $\frac{1}{\sqrt{2}} = -3dB$  of steady-state value.

• Proper transfer function  $\stackrel{\triangle}{=} \lim_{\omega \to \infty} |G(j\omega)| < \infty$ . Strictly proper transfer function  $\stackrel{\triangle}{=} \lim_{\omega \to \infty} |G(j\omega)| = 0$ .

First-order Systems : a Second Look

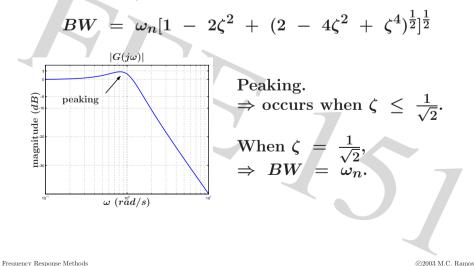
- Bandwidth BW = |a|. Since  $\tau = \frac{1}{a}$ ,
  - $\Rightarrow more bandwidth = faster system$ less bandwidth = slower system
- However, more bandwidth usually leads to more noise getting through the system.





• Second-order system.

• First-order system.



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- Easy to generate sinusoids / test signals.
- Straightforward to replace s with  $j\omega$  in the transfer function to get  $G(j\omega)$ .
- The plot of  $|G(j\omega)|$  and  $\angle G(j\omega)$  gives an insight about the system for analysis and design.

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## **Polar Plots**

• Polar plots - plot  $G(j\omega)$  on the complex plane as  $\omega$  is varied from 0 to  $\infty$  (or  $-\infty$  to  $\infty$ ).

$$egin{array}{rll} G(j\omega) &=& G(s)|_{s=j\omega} &=& \underbrace{R(\omega)}_{\mathrm{real}} &+& j \underbrace{X(\omega)}_{\mathrm{imag}} \ G(j\omega) &=& |G(j\omega)| \angle \phi(\omega) \end{array}$$

where

$$egin{aligned} |G(j\omega)|^2 &= [R(j\omega)]^2 + [X(\omega)]^2 \ \phi(\omega) &= ext{tan}^{-1} \left[rac{X(j\omega)}{R(\omega)}
ight] \end{aligned}$$

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• Frequency domain  $\leftrightarrow$  time domain connection.  $\Rightarrow$  sometimes difficult to see.

$$F(s) = \mathscr{L}{f(t)} = \int_0^\infty f(t)e^{-st}dt$$

$$f(t) = \mathscr{L}^{-1}{F(s)} = \frac{1}{j2\pi} \int_{\sigma-j\infty}^{\sigma+j\infty} F(s)e^{st}ds$$

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Polar Plots

• Example. RC filter.

$$G(s) \;=\; rac{1}{RCs\;+\;1}$$

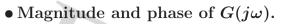
Determining 
$$G(j\omega)$$
.  

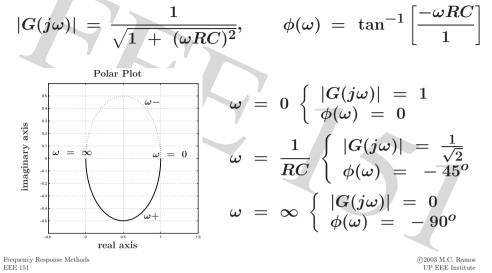
$$G(j\omega) = \frac{1}{j\omega RC + 1} \cdot \frac{j\omega RC - 1}{j\omega RC - 1}$$

$$= \frac{-1}{-(\omega RC)^2 - 1} + \frac{j\omega RC}{-(\omega RC)^2 - 1}$$

$$= \frac{1}{1 + (\omega RC)^2} + \frac{-j\omega RC}{1 + (\omega RC)^2}$$

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- Summary
- What do we mean by frequency response?

Basically, we look at the steady-state response to a sinusoidal input.

- We usually talk about magnitude and phase when dealing with frequency response.
- We must get reacquainted with *n*th-order systems in terms of frequency response characteristics.
- Polar plots. Up next : Bode plots, at better way of looking at things.

- Simple graphical presentation of the transfer function. Only one plot describes the transfer function completely.
- Calculations are tedious and frequent recalculations are required for system design.
- Does not show the direct effect of poles and zeros to the system response.

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