# Frequency Response

- Response to a step input is a primary consideration in discussions on control systems.
- What about responses to other types of input? Consider a sinusoidal input.
- For LTI systems, the response to a sinusoidal input is a sinusoid with the same frequency as the input sinusoid.

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## Frequency Response

- Develop techniques and tools for system design in the frequency domain.
  - -Polar plots.
  - -Bode plots.
  - -Nyquist criterion and Nyquist plot.
- For today, we will see what frequency response is all about. Why do we need to use frequency response techniques?

We will look at a simple example of a polar plot.

### Frequency Response

- However, the sinusodial output has a different magnitude and phase in comparison to the sinusoidal input.
- Determine the magnitude and phase variations of the sinusoidal response when the frequency is varied.
- Define frequency domain specifications.

  Relate time domain performance specifications to the frequency domain specifications.

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## Frequency Response

• What is the frequency response?

Frequency response - steady-state response of a system to a sinusoidal input signal.



• Consider a system G(s) with unique poles  $p_1, \ldots, p_n$ .

$$G(s) = \frac{(s - z_1) \dots (s - z_m)}{(s - p_1) \dots (s - p_n)}$$

• Let the input  $r(t) = A \sin \omega t$ .

$$R(s) = \frac{A\omega}{s^2 + \omega^2}$$

• The output is Y(s) = G(s)R(s).

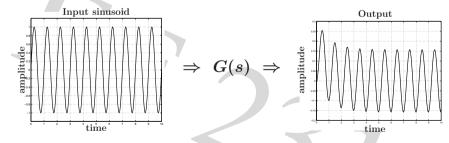
$$Y(s) = rac{k_1}{s - p_1} + \dots + rac{k_n}{s - p_n} + rac{lpha s + eta}{s^2 + \omega^2}$$
 $y(t) = k_1 e^{p_1 t} + \dots + k_n e^{p_n t} + \mathscr{L}^{-1} \left\{ rac{lpha s + eta}{s^2 + \omega^2} 
ight\}$ 

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# Detailed Look at the Frequency Response

• What does the actual response look like.



• Response to the sinusoid is composed of a transient part and steady-state part.

$$y(t) = y_{transient}(t) + y_{steady-state}(t)$$

• Using the final value theorem.

(assuming poles  $p_1, \ldots, p_n$  are in the LHP),

$$\lim_{t \to \infty} y(t) = \lim_{t \to \infty} \mathcal{L}^{-1} \left\{ \frac{\alpha s + \beta}{s^2 + \omega^2} \right\}$$

The steady-state output y(t) as  $t \to \infty$ ,

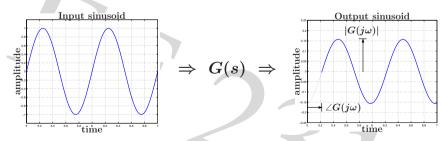
$$egin{aligned} y(t) &= rac{1}{\omega} |A\omega G(j\omega)| \sin(\omega t \ + \ \phi) \ &= A|G(j\omega)| \sin(\omega t \ + \ \phi) \qquad ext{where } \phi \ = \ \angle G(j\omega) \end{aligned}$$

• Output depends on the magnitude and phase of  $G(j\omega)$ .

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# Detailed Look at the Frequency Response

• Steady-state response.



• For the input signal  $\sin(\omega t)$ , the output has a frequency  $\omega$ , a peak value of  $|G(j\omega)|$  and is shifted by  $\angle G(j\omega)$  with respect to the input.

# Detailed Look at the Frequency Response

ullet Magnitude  $|G(j\omega)|$  expressed in dB (decibels). Logarithmic gain  $\stackrel{\triangle}{=} 20 \log |G(j\omega)|$ .

- Bandwidth  $\stackrel{\triangle}{=}$  frequency where magnitude drops by  $\frac{1}{\sqrt{2}} = -3dB$  of steady-state value.
- ullet Proper transfer function  $\stackrel{\triangle}{=} \lim_{\omega o \infty} |G(j\omega)| < \infty.$  Strictly proper transfer function  $\stackrel{\triangle}{=} \lim_{\omega o \infty} |G(j\omega)| = 0.$

## First-order Systems : a Second Look

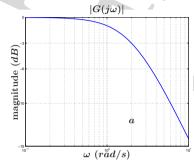
- Bandwidth BW = |a|. Since  $\tau = \frac{1}{a}$ ,

  ⇒ more bandwidth = faster system
  less bandwidth = slower system
- However, more bandwidth usually leads to more noise getting through the system.

## First-order Systems: a Second Look

• First-order system.

$$G(s) = \frac{a}{s+a} = \frac{1}{\tau s+1}$$
 where  $\tau = \frac{1}{a}$ 



Lower output at higher frequencies ( $\omega > a$ ).

High frequencies (e.g. noise) are filtered.

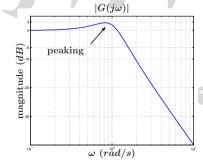
- $\Rightarrow$  doubling = +6dB.
- $\Rightarrow$  halving = -6dB.

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# Second-order Systems: a Second Look

ullet Second-order system.

$$BW = \omega_n [1 - 2\zeta^2 + (2 - 4\zeta^2 + \zeta^4)^{\frac{1}{2}}]^{\frac{1}{2}}$$



Peaking.

 $\Rightarrow$  occurs when  $\zeta \leq \frac{1}{\sqrt{2}}$ .

When 
$$\zeta = \frac{1}{\sqrt{2}}$$
,  $\Rightarrow BW = \omega_n$ .

- Easy to generate sinusoids / test signals.
- Straightforward to replace s with  $j\omega$  in the transfer function to get  $G(j\omega)$ .
- The plot of  $|G(j\omega)|$  and  $\angle G(j\omega)$  gives an insight about the system for analysis and design.

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## Polar Plots

• Polar plots - plot  $G(j\omega)$  on the complex plane as  $\omega$  is varied from 0 to  $\infty$  (or  $-\infty$  to  $\infty$ ).

$$G(j\omega) = G(s)|_{s=j\omega} = \underbrace{R(\omega)}_{\mathrm{real}} + j\underbrace{X(\omega)}_{\mathrm{imag}}$$
 $G(j\omega) = |G(j\omega)| \angle \phi(\omega)$ 

where

$$|G(j\omega)|^2 = [R(j\omega)]^2 + [X(\omega)]^2$$
  
 $\phi(\omega) = \tan^{-1} \left[\frac{X(j\omega)}{R(\omega)}\right]$ 

◆ Frequency domain ↔ time domain connection.
⇒ sometimes difficult to see.

$$F(s) = \mathscr{L}{f(t)} = \int_0^\infty f(t)e^{-st}dt$$
 
$$f(t) = \mathscr{L}^{-1}{F(s)} = \frac{1}{j2\pi} \int_{\sigma-j\infty}^{\sigma+j\infty} F(s)e^{st}ds$$

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#### **Polar Plots**

• Example. RC filter.

$$G(s) = \frac{1}{RCs + 1}$$

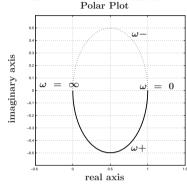
• Determining  $G(j\omega)$ .

$$G(j\omega) = rac{1}{j\omega RC + 1} \cdot rac{j\omega RC - 1}{j\omega RC - 1}$$
 $= rac{-1}{-(\omega RC)^2 - 1} + rac{j\omega RC}{-(\omega RC)^2 - 1}$ 
 $= rac{1}{1 + (\omega RC)^2} + rac{-j\omega RC}{1 + (\omega RC)^2}$ 

#### **Polar Plots**

• Magnitude and phase of  $G(j\omega)$ .

$$|G(j\omega)| \, = \, rac{1}{\sqrt{1 \, + \, (\omega RC)^2}}, \qquad \phi(\omega) \, = \, an^{-1} \left[rac{-\omega RC}{1}
ight]$$



$$\omega = 0 \begin{cases} |G(j\omega)| = 1\\ \phi(\omega) = 0 \end{cases}$$

$$\omega = \frac{1}{RC} \begin{cases} |G(j\omega)| = \frac{1}{\sqrt{2}}\\ \phi(\omega) = -45^{o} \end{cases}$$

 $\omega ~=~ \infty ~\left\{egin{array}{l} |G(j\omega)| &=~ 0 \ \phi(\omega) &=~ -~ 90^o \end{array}
ight.$ 

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# Summary

- What do we mean by frequency response?

  Basically, we look at the steady-state response to a sinusoidal input.
- We usually talk about magnitude and phase when dealing with frequency response.
- We must get reacquainted with nth-order systems in terms of frequency response characteristics.
- Polar plots. Up next: Bode plots, at better way of looking at things.

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#### Remarks on Polar Plots

• Simple graphical presentation of the transfer function.

Only one plot describes the transfer function completely.

• Calculations are tedious and frequent recalculations are required for system design.

• Does not show the direct effect of poles and zeros to the system response.

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